

## 4 X 45W QUAD BRIDGE CAR RADIO AMPLIFIER PLUS HSD

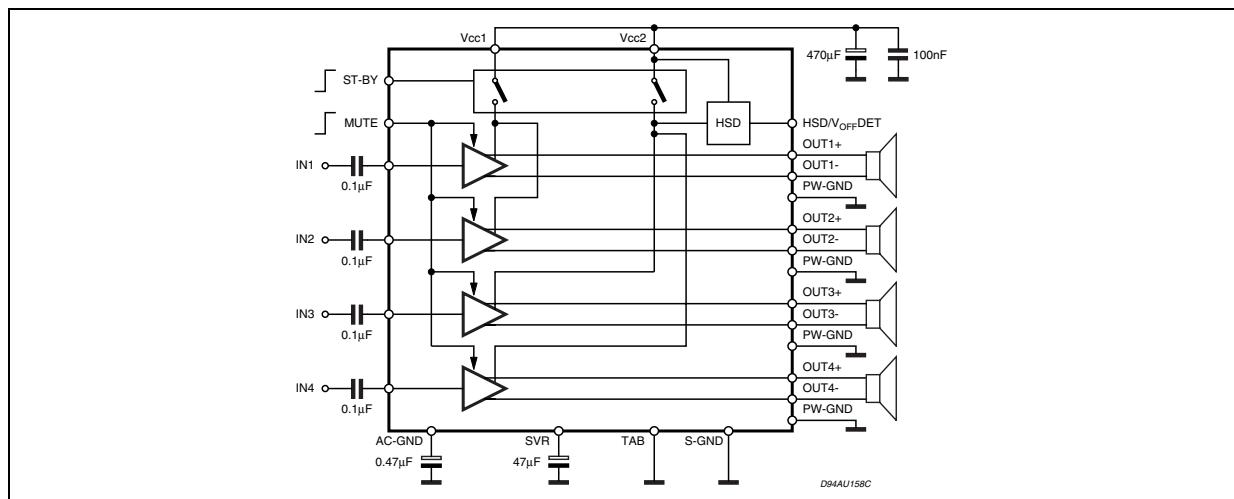
### 1 Features

- SUPERIOR OUTPUT POWER CAPABILITY:
  - 4 x 50W/4Ω MAX.
  - 4 x 45W/4Ω EIAJ
  - 4 x 30W/4Ω @ 14.4V, 1KHz, 10%
  - 4 x 80W/2Ω MAX.
  - 4 x 77W/2Ω EIAJ
  - 4 x 55W/2Ω @ 14.4V, 1KHz, 10%
- MULTIPOWER BCD TECHNOLOGY
- MOSFET OUTPUT POWER STAGE
- EXCELLENT 2Ω DRIVING CAPABILITY
- HI-FI CLASS DISTORTION
- LOW OUTPUT NOISE
- ST-BY FUNCTION
- MUTE FUNCTION
- AUTOMUTE AT MIN. SUPPLY VOLTAGE DETECTION
- LOW EXTERNAL COMPONENT COUNT:
  - INTERNALLY FIXED GAIN (26dB)
  - NO EXTERNAL COMPENSATION
  - NO BOOTSTRAP CAPACITORS
- ON BOARD 0.35A HIGH SIDE DRIVER

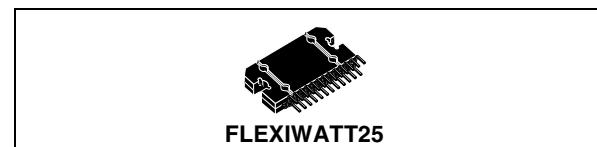
#### 1.1 Protections:

- OUTPUT SHORT CIRCUIT TO GND, TO VS, ACROSS THE LOAD
- VERY INDUCTIVE LOADS
- OVERRATING CHIP TEMPERATURE WITH SOFT THERMAL LIMITER

**Figure 2. Block Diagram**



**Figure 1. Package**



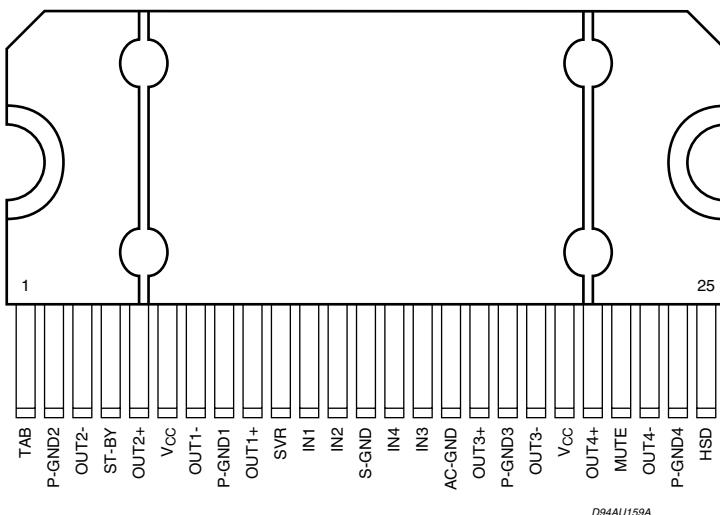
**Table 1. Order Codes**

Part Number	Package
TDA7560	FLEXIWATT25

- OUTPUT DC OFFSET DETECTION
- LOAD DUMP VOLTAGE
- FORTUITOUS OPEN GND
- REVERSED BATTERY
- ESD

### 2 Description

The TDA7560 is a breakthrough BCD (Bipolar / CMOS / DMOS) technology class AB Audio Power Amplifier in Flexiwatt 25 package designed for high power car radio. The fully complementary P-Channel/N-Channel output structure allows a rail to rail output voltage swing which, combined with high output current and minimised saturation losses sets new power references in the car-radio field, with unparalleled distortion performances.

**Figure 3. Pin Connection (Top view)****Table 2. Absolute Maximum Ratings**

Symbol	Parameter	Value	Unit
$V_{CC}$	Operating Supply Voltage	18	V
$V_{CC}$ (DC)	DC Supply Voltage	28	V
$V_{CC}$ (pk)	Peak Supply Voltage (for $t = 50\text{ms}$ )	50	V
$I_O$	Output Peak Current Repetitive (Duty Cycle 10% at $f = 10\text{Hz}$ ) Non repetitive ( $t = 100\mu\text{s}$ )	9 10	A A
$P_{tot}$	Power Dissipation $T_{case} = 70^\circ\text{C}$	80	W
$T_j$	Junction Temperature	150	$^\circ\text{C}$
$T_{stg}$	Storage Temperature	-55 to 150	$^\circ\text{C}$

**THERMAL DATA**

Symbol	Parameter	Value	Unit
$R_{th j-case}$	Thermal Resistance Junction to case	Max.	1 $^\circ\text{C/W}$

**Table 3. Electrical Characteristics**

(Refer to the test and application diagram,  $V_S = 13.2V$ ;  $R_L = 4\Omega$ ;  $R_g = 600\Omega$ ;  $f = 1KHz$ ;  $T_{amb} = 25^\circ C$ ; unless otherwise specified).

Symbol	Parameter	Test Condition	Min.	Typ.	Max.	Unit
$I_{q1}$	Quiescent Current	$R_L = \infty$	120	200	320	mA
$V_{os}$	Output Offset Voltage	Play Mode			$\pm 60$	mV
$dV_{os}$	During mute ON/OFF output offset voltage				$\pm 60$	mV
$G_V$	Voltage Gain		25	26	27	dB
$dG_V$	Channel Gain Unbalance				$\pm 1$	dB
$P_o$	Output Power	$V_S = 13.2V$ ; THD = 10%	23	25		W
		$V_S = 13.2V$ ; THD = 1%	16	19		W
		$V_S = 14.4V$ ; THD = 10%	28	30		W
		$V_S = 14.4V$ ; THD = 1%	20	23		W
		$V_S = 13.2V$ ; THD = 10%, $2\Omega$	42	45		W
		$V_S = 13.2V$ ; THD = 1%, $2\Omega$	32	34		W
		$V_S = 14.4V$ ; THD = 10%, $2\Omega$	50	55		W
		$V_S = 14.4V$ ; THD = 1%, $2\Omega$	40	43		W
$P_{o EIAJ}$	EIAJ Output Power (*)	$V_S = 13.7V$ ; $R_L = 4\Omega$ $V_S = 13.7V$ ; $R_L = 2\Omega$	41 72	45 77		W W
$P_{o max.}$	Max. Output Power (*)	$V_S = 14.4V$ ; $R_L = 4\Omega$ $V_S = 14.4V$ ; $R_L = 2\Omega$		50 80		W W
THD	Distortion	$P_o = 4W$ $P_o = 15W$ ; $R_L = 2\Omega$		0.006 0.015	0.05 0.07	% %
eNo	Output Noise	"A" Weighted $Bw = 20Hz$ to $20Khz$		35 50	50 70	$\mu V$ $\mu V$
SVR	Supply Voltage Rejection	$f = 100Hz$ ; $V_r = 1Vrms$	50	70		dB
$f_{ch}$	High Cut-Off Frequency	$P_O = 0.5W$	100	300		KHz
$R_i$	Input Impedance		80	100	120	$K\Omega$
$C_T$	Cross Talk	$f = 1Khz P_O = 4W$ $f = 10Khz P_O = 4W$	60	70 60	- -	dB dB
$I_{SB}$	St-By Current Consumption	$V_{St-By} = 1.5V$			20	$\mu A$
$I_{pin5}$	St-by pin Current	$V_{St-By} = 1.5V$ to $3.5V$			$\pm 10$	$\mu A$
$V_{SB\ out}$	St-By Out Threshold Voltage	(Amp: ON)	3.5			V
$V_{SB\ in}$	St-By In Threshold Voltage	(Amp: OFF)			1.5	V
$A_M$	Mute Attenuation	$P_{Oref} = 4W$	80	90		dB
$V_M\ out$	Mute Out Threshold Voltage	(Amp: Play)	3.5			V
$V_M\ in$	Mute In Threshold Voltage	(Amp: Mute)			1.5	V
$V_{AM\ in}$	VS Automute Threshold	(Amp: Mute) $Att \geq 80dB$ ; $P_{Oref} = 4W$	6.5	7		V
		(Amp: Play) $Att < 0.1dB$ ; $P_O = 0.5W$		7.5	8	V
$I_{pin23}$	Muting Pin Current	$V_{MUTE} = 1.5V$ (Sourced Current)	7	12	18	$\mu A$
		$V_{MUTE} = 3.5V$	-5		18	$\mu A$
<b>HSD SECTION</b>						
$V_{dropout}$	Dropout Voltage	$I_O = 0.35A$ ; $V_S = 9$ to $16V$		0.25	0.6	V
$I_{prot}$	Current Limits		400		800	mA

**Table 3. Electrical Characteristics** (continued)

(Refer to the test and application diagram,  $V_S = 13.2V$ ;  $R_L = 4\Omega$ ;  $R_g = 600\Omega$ ;  $f = 1KHz$ ;  $T_{amb} = 25^\circ C$ ; unless otherwise specified).

Symbol	Parameter	Test Condition	Min.	Typ.	Max.	Unit
<b>OFFSET DETECTOR (Pin 26)</b>						
$V_{M\_ON}$	Mute Voltage for DC offset detection enabled	$V_{stby} = 5V$	8			V
$V_{M\_OFF}$				6		V
$V_{OFF}$	Detected Differential Output Offset	$V_{stby} = 5V$ ; $V_{mute} = 8V$	$\pm 2$	$\pm 3$	$\pm 4$	V
$V_{25\_T}$	Pin 25 Voltage for Detection = TRUE	$V_{stby} = 5V$ ; $V_{mute} = 8V$ $V_{OFF} > \pm 4V$	0		1.5	V
$V_{25\_F}$	Pin 25 Voltage for Detection = FALSE	$V_{stby} = 5V$ ; $V_{mute} = 8V$ $V_{OFF} > \pm 2V$	12			V

(\*) Saturated square wave output.

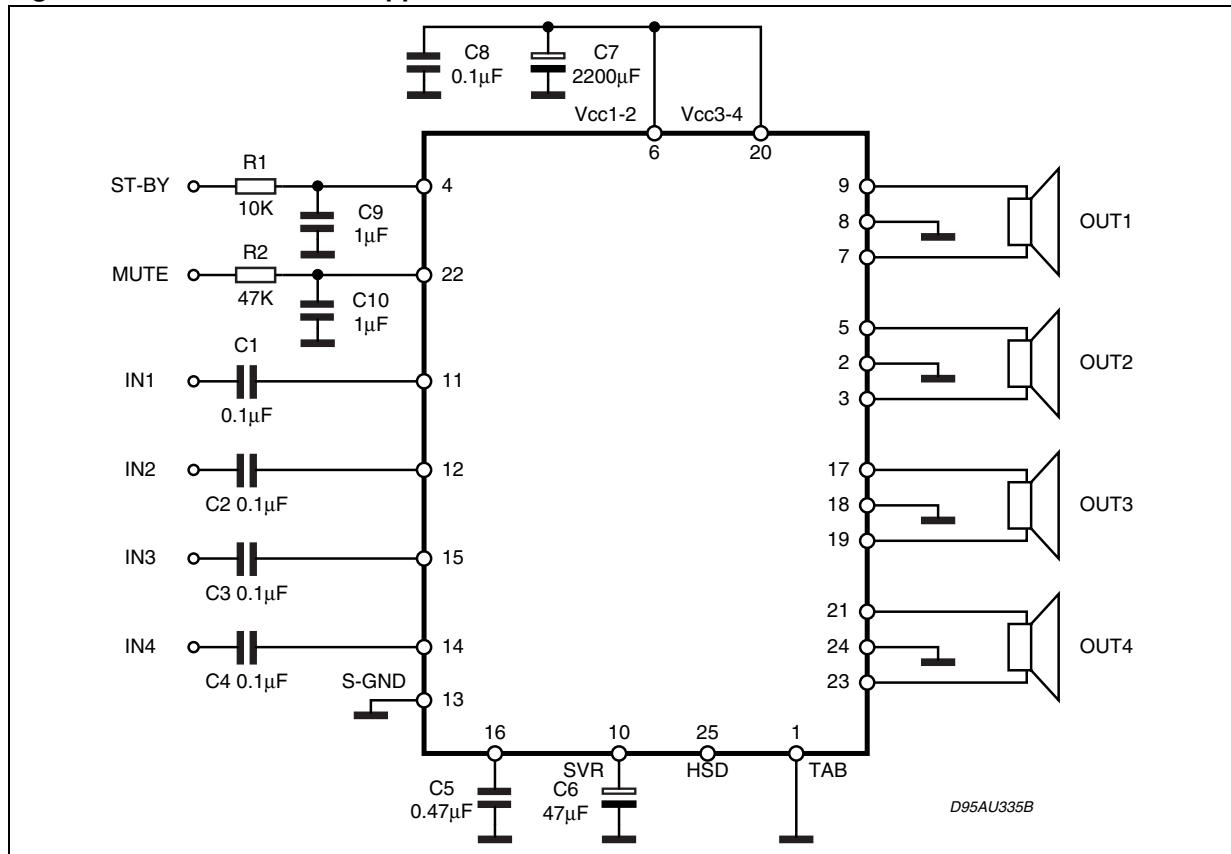
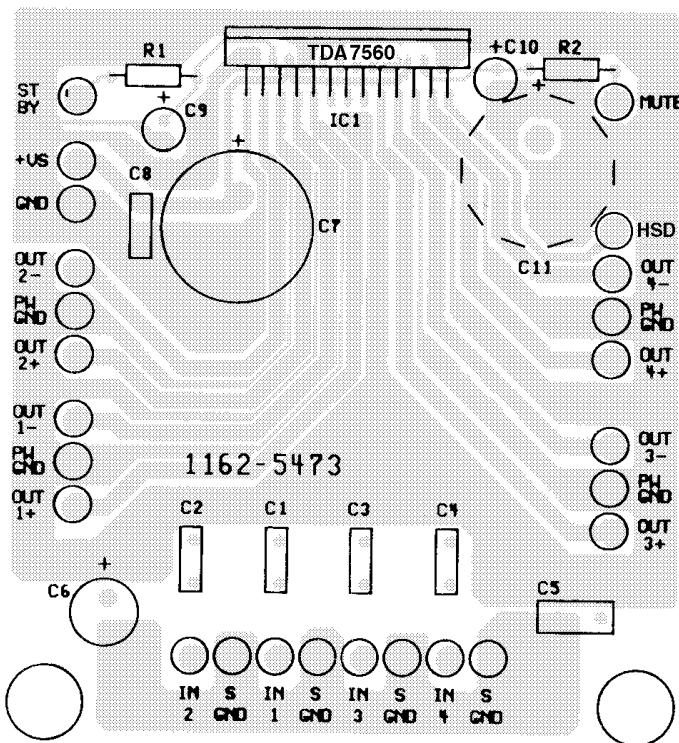
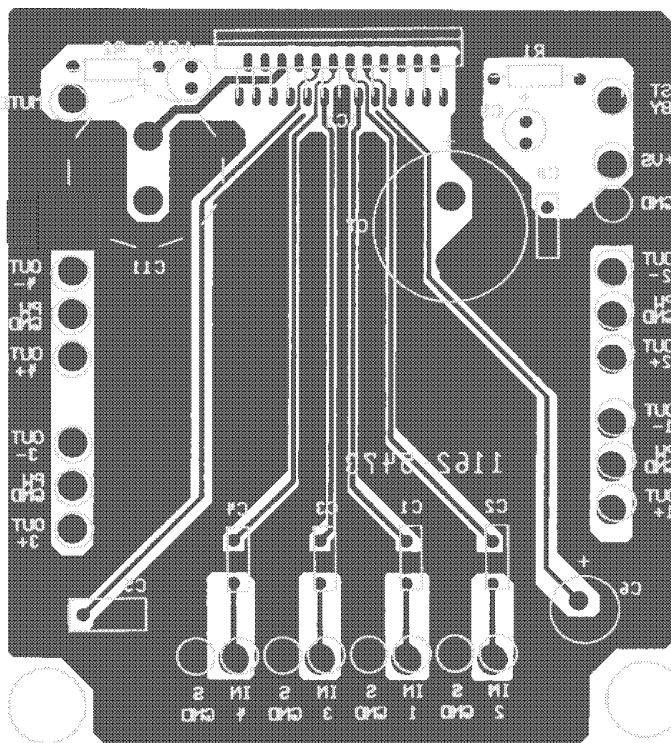
**Figure 4. Standard Test and Application Circuit**

Figure 5. P.C.B. and component layout of the Figure 4.

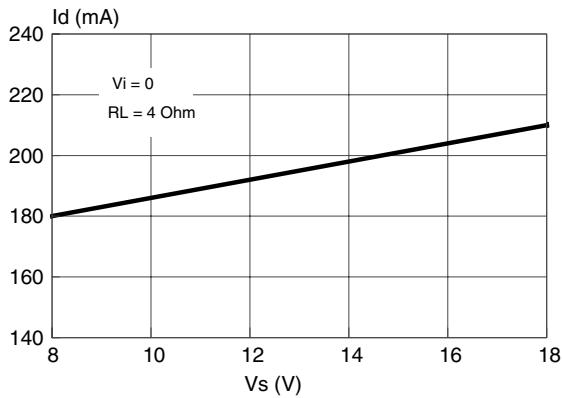
### Components & Top Copper Layer



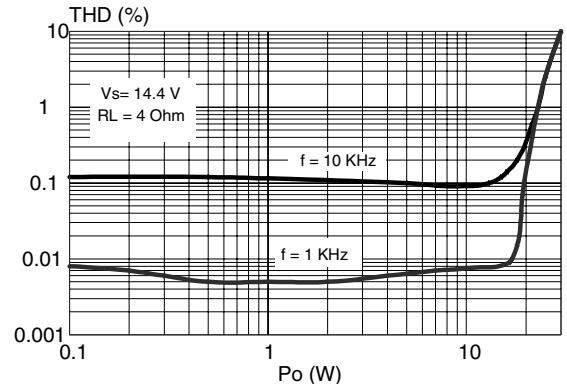
### Bottom Copper Layer



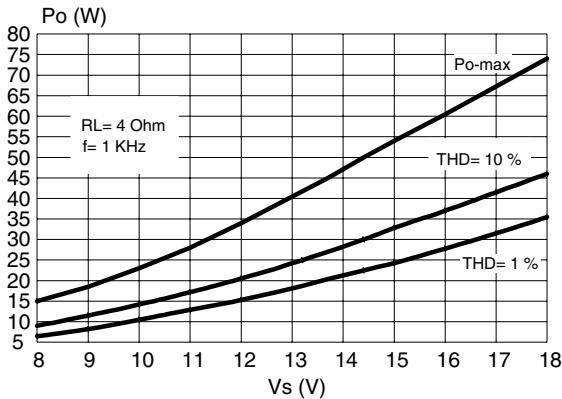
**Figure 6. Quiescent current vs. supply voltage.**



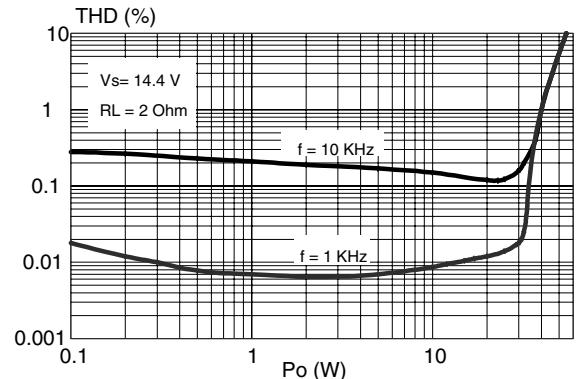
**Figure 9. Distortion vs. output Power**



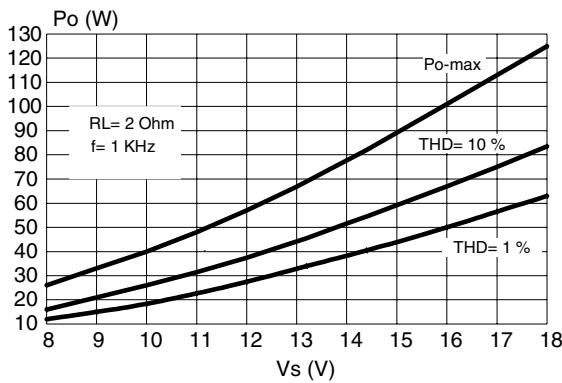
**Figure 7. Output power vs. supply voltage.**



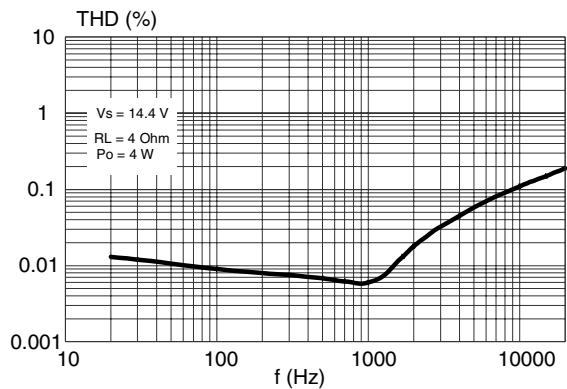
**Figure 10. Distortion vs. output power**

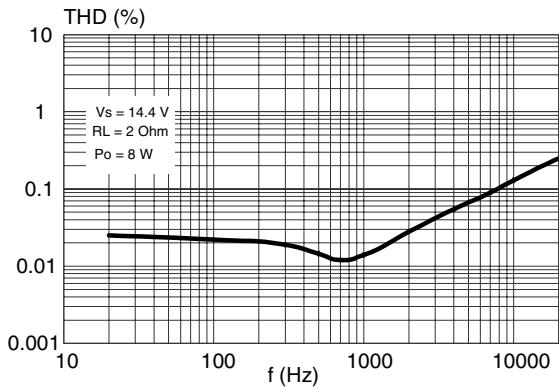
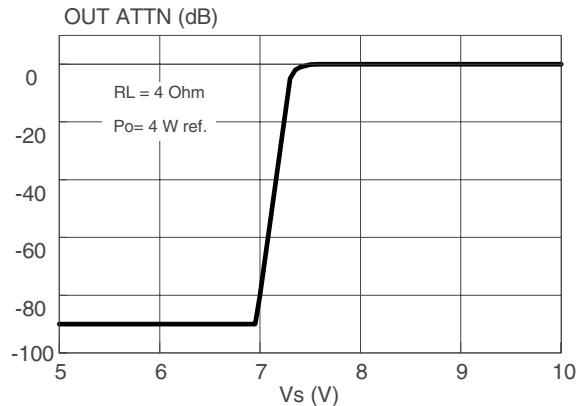
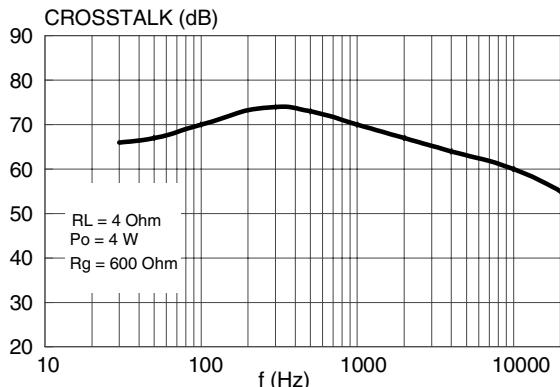
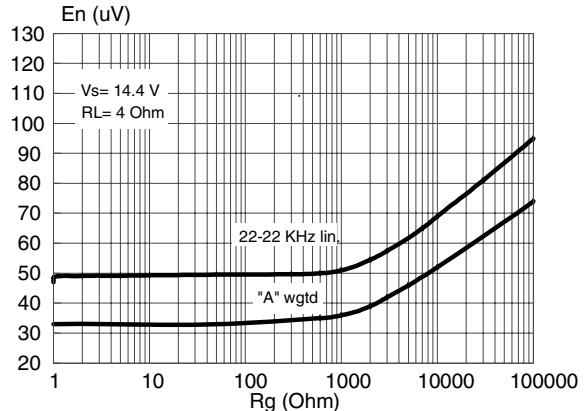
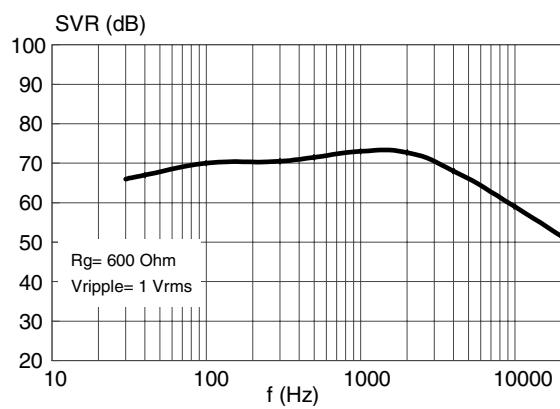
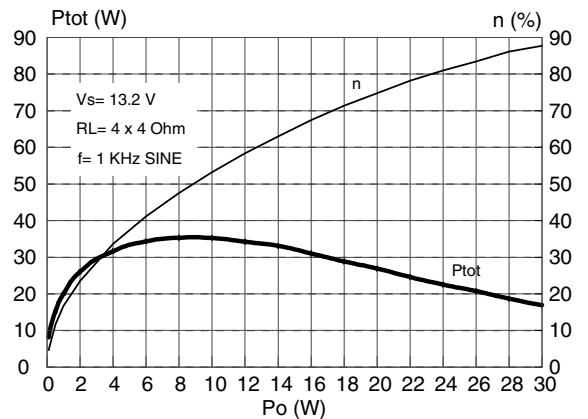


**Figure 8. Output power vs. supply voltage.**

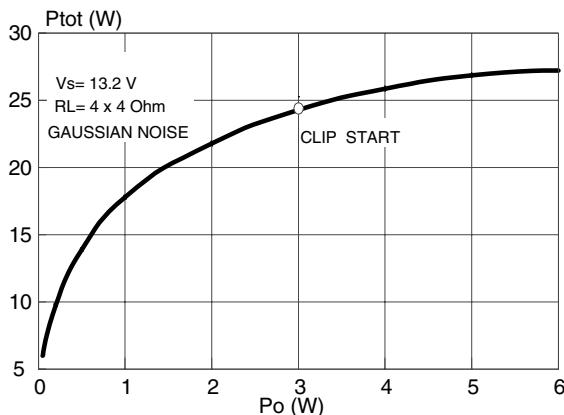


**Figure 11. Distortion vs. frequency.**

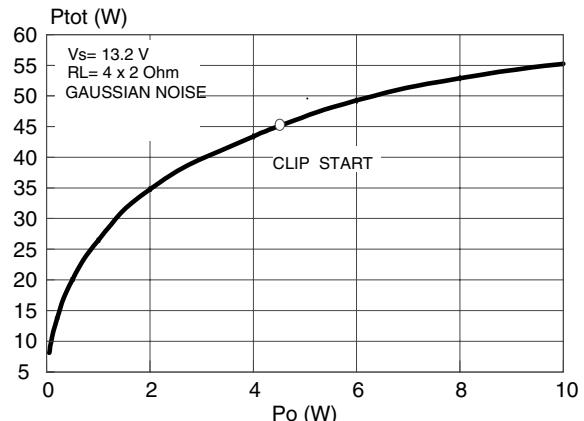


**Figure 12. Distortion vs. frequency.****Figure 15. Output attenuation vs. supply volt.****Figure 13. Crosstalk vs. frequency.****Figure 16. Output noise vs. source resistance.****Figure 14. Supply voltage rejection vs. freq.****Figure 17. Power dissipation & efficiency vs. output power (sine-wave operation)**

**Figure 18. Power dissipation vs. output power (Music/Speech Simulation)**



**Figure 19. Power dissipation vs. output power (Music/Speech Simulation)**



### 3 DC Offset Detector

The TDA7560 integrates a DC offset detector to avoid that an anomalous DC offset on the inputs of the amplifier may be multiplied by the gain and result in a dangerous large offset on the outputs which may lead to speakers damage for overheating.

The feature is enabled by the MUTE pin and works with the amplifier muted and with no signal on the inputs. The DC offset detection is signaled out on the HSD pin.

### 4 Application Hints (ref. to the circuit of fig. 4)

#### 4.1 SVR

Besides its contribution to the ripple rejection, the SVR capacitor governs the turn ON/OFF time sequence and, consequently, plays an essential role in the pop optimization during ON/OFF transients. To conveniently serve both needs, **ITS MINIMUM RECOMMENDED VALUE IS 10 $\mu$ F**.

#### 4.2 INPUT STAGE

The TDA7560's inputs are ground-compatible and can stand very high input signals ( $\pm 8V_{pk}$ ) without any performances degradation.

If the standard value for the input capacitors (0.1 $\mu$ F) is adopted, the low frequency cut-off will amount to 16 Hz.

#### 4.3 STAND-BY AND MUTING

STAND-BY and MUTING facilities are both CMOS-COMPATIBLE. In absence of true CMOS ports or microprocessors, a direct connection to Vs of these two pins is admissible but a 470 kOhm equivalent resistance should present between the power supply and the muting and stand-by pins.

R-C cells have always to be used in order to smooth down the transitions for preventing any audible transient noises.

About the stand-by, the time constant to be assigned in order to obtain a virtually pop-free transition has to be slower than 2.5V/ms.

#### 4.4 HEATSINK DEFINITION

Under normal usage (4 Ohm speakers) the heatsink's thermal requirements have to be deduced from fig. 18, which reports the simulated power dissipation when real music/speech programmes are played out. Noise with gaussian-distributed amplitude was employed for this simulation. Based on that, frequent clipping occurrence (worst-case) will cause  $P_{diss} = 26W$ . Assuming  $T_{amb} = 70^{\circ}C$  and  $T_{CHIP} = 150^{\circ}C$  as boundary conditions, the heatsink's thermal resistance should be approximately 2°C/W. This would avoid any thermal shutdown occurrence even after long-term and full-volume operation.

## 5 Package Information

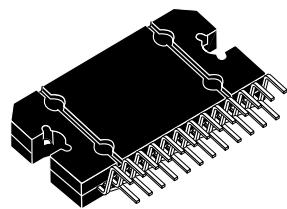
Figure 20. Flexiwatt25 (vertical) Mechanical Data & Package Dimensions

DIM.	mm			inch		
	MIN.	Typ.	MAX.	MIN.	Typ.	MAX.
A	4.45	4.50	4.65	0.175	0.177	0.183
B	1.80	1.90	2.00	0.070	0.074	0.079
C		1.40			0.055	
D	0.75	0.90	1.05	0.029	0.035	0.041
E	0.37	0.39	0.42	0.014	0.015	0.016
F (1)			0.57			0.022
G	0.80	1.00	1.20	0.031	0.040	0.047
G1	23.75	24.00	24.25	0.935	0.945	0.955
H (2)	28.90	29.23	29.30	1.139	1.150	1.153
H1		17.00			0.669	
H2		12.80			0.503	
H3		0.80			0.031	
L (2)	22.07	22.47	22.87	0.869	0.884	0.904
L1	18.57	18.97	19.37	0.731	0.747	0.762
L2 (2)	15.50	15.70	15.90	0.610	0.618	0.626
L3	7.70	7.85	7.95	0.303	0.309	0.313
L4		5			0.197	
L5		3.5			0.138	
M	3.70	4.00	4.30	0.145	0.157	0.169
M1	3.60	4.00	4.40	0.142	0.157	0.173
N	2.20				0.086	
O	2				0.079	
R	1.70				0.067	
R1	0.5				0.02	
R2	0.3				0.12	
R3	1.25				0.049	
R4	0.50				0.019	
V			5° (T p.)			
V1			3° (Typ.)			
V2			20° (Typ.)			
V3			45° (Typ.)			

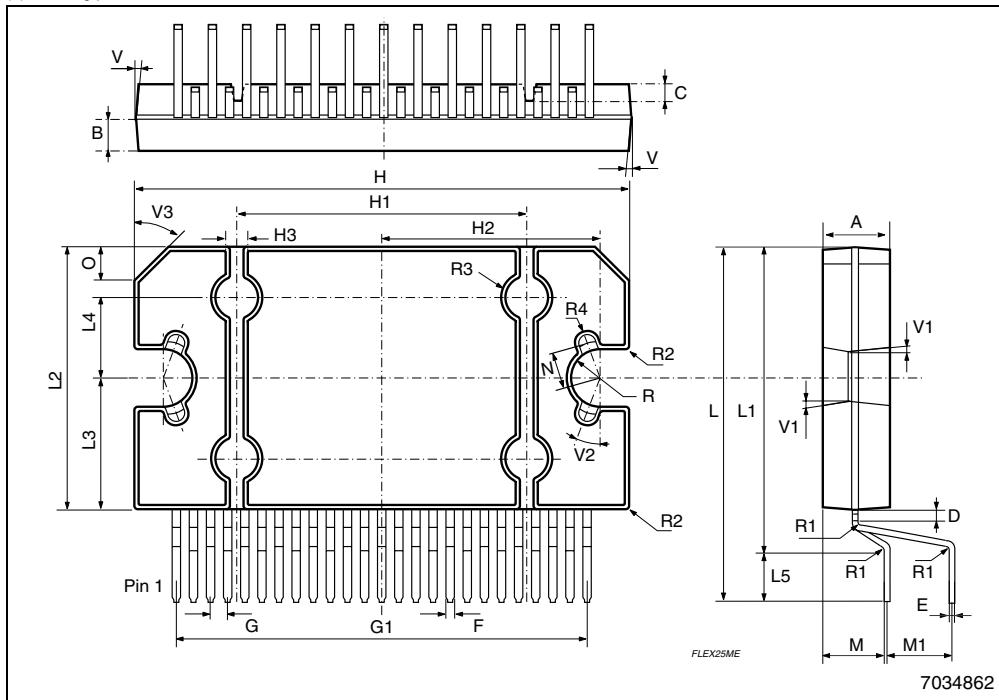
(1): dam-bar protrusion not included

(2): molding protrusion included

### OUTLINE AND MECHANICAL DATA



Flexiwatt25 (vertical)



## 6 Revision History

**Table 4. Revision History**

Date	Revision	Description of Changes
December 2001	1	First Issue
February 2005	2	Improved value from 75 to 20µA of the “ST_BY Current Consumption” parameter in the table 3 at the page 3.

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